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Abstract—In earlier work, we suggested a low cost control communication scheme across half-duplex Ethernet allowing for reduced jitter and delay for an information sharing control communication bus. The principle advantage of half-duplex Ethernet is the reduced number of wires (two versus four wires) and the reduced cost of switch/router equipment. The basic principles for our communication has been the network wide introduction of a synchronisation signal and the modification of the back-off scheme for each node, by assigning to each a node specific minimal back-off time. In this paper, a Markov chain model for modeling network jitter and delays is developed for this control communication approach. For this, the events occurring after a synchronization signal are formally summarized within a state vector. A two node example is considered. The state vector describes the states of collision detection, back-off and transmission within the two node system. Relevant probabilities are assigned for transfer between different states. The process of calculating the overall Markov chain model and related parameters is described. This approach shows the advantage of our approach over a standard Carrier Sense Multiple Access/Collision Detection (CSMA/CD) setting.

Keywords—Markov Model, Backoff time, Jitter, Delays, Ethernet, and network model.

I. INTRODUCTION

For the last few decades, the increasing usage of electronics and computerised systems significantly impacted on the industry, automotive systems, avionics and robotics. In order to achieve a given task, these engineering systems will share sensor and actuator information in a distributed sense, e.g. torque, velocity, current and position. In this scenario, the communication between the technological systems and its sub-units is one of the main factors which contribute to the capabilities and effectiveness. Hence, the applications of embedded systems should timely deliver synchronized data-sets, minimize latency in their response and meet their performance target. Networked control communication is a backbone to accomplish tasks to be carried out for distributed dynamic control using embedded systems. With that, there are various types of communication protocols such as Controller Area Network (CAN), CANopen, Flexray, Time Triggered CAN (TTCAN) and Time Triggered Ethernet (TTEthernet) [1], [2], [3], [4].

Modern wired, high speed Ethernet based systems (Profinet [5], [6], EtherCAT [7], [8] and Avionics Full Duplex Switched Ethernet (AFDX) [9]) have been very desirable due to the high data bandwidth of up to 1 GB/s and its reliability when operated as a switched, full-duplex network. However, such configuration can be for vehicular producers disadvantageous due to the increased weight of a four wire (two twisted pair) configuration and the added need for specific switches to create such network. A solution to this problem has been recently created, BroadR-Reach [10], which is again a switched network and full-duplex but using only two wires, i.e. a single twisted pair.

Despite this, within recent years, we have been interested in a control network solution using half-duplex Ethernet based on the two wire solution [11], [12]. For this reason, we investigated and developed Media Access Control (MAC)-strategies for half-duplex Ethernet which allow to minimize jitter and delay for control communication. Such strategies are easily implemented on existing Ethernet systems. Using them in the half-duplex frame work reduces weight and cost for switch/router technology (as it can be in principle reduced). In comparison to recent solutions [10], our solution [11], [12] allows complete information sharing across the network (such as CAN) and is rather low cost.

Our approach has been inspired by the CANOpen protocol, where a synchronisation signal triggers nodes connected to the network to send their control signal data packet. In case, several nodes send a data packet, the CSMA/CA-mechanism (Carrier Sense Multiple Access/Collision Avoidance) of the CAN protocol will resolve the order of transmission of these data packets. The synchronisation signal combined with the CDMA/CA scheme guarantees data arrival with minimal jitter and delay. Such a scheme within half-duplex Ethernet is not possible due to the inherent CDMA/CD MAC-strategy. However, a global synchronization signal is easily implemented using for instance IEEE 1588 clock synchronization. Thus, it is
in the second step necessary to review and modify the Ethernet MAC (Binary Exponential Backoff, BEB) to avoid packet collisions as much as possible so that transmission delay and jitter following the synchronization signal is minimal. This was achieved in the practical investigations of [11], [12] by providing each transmission node with a specific minimal back-off time. Thus, the basic principle of the suggested communication scheme is to have all nodes sending at a synchronization event, for which then the back-off scheme with specific minimal back-off times minimizes jitter and delay. It was shown in [11], [12] using simulations and practical tests that jitter and delay can be practically kept smaller than with a standard BEB scheme and equal minimal back-off time.

In this paper, a mathematical model is developed for the suggested communication scheme, i.e. this is to provide theoretical evidence for the positive effect of reducing jitter and delay using the suggested MAC. For this reason, a Markov model [13] is derived considering the different states of the back-off scheme, successful transmission, backoff or packet collision. By modelling every single state using a computational approach, it is possible to obtain an overall value for jitter and delay. This is done here for a two node system as an example. Moreover, the back-off scheme is investigated in the context of the suggested communication principles for both, Binary Exponential Backoff and Linear Backoff. Note that for simplicity the network is modelled without any delays in the transmission wiring.

In the next Section, the basic Binary Exponential Back-off scheme is revisited to allow for a better understanding of the suggested control communication approach.

II. HALF DUPLEX ETHERNET REVISED

Before going into the detail of our MAC for control communication we review the basic half duplex Ethernet MAC ideas.

A. Half Duplex Ethernet MAC

The MAC layer of the IEEE 802.3 standard for Ethernet uses the CSMA/CD (Carrier Sense Multiple Access/Collision Detection) protocol with Binary Exponential Backoff [14], [15]. At a time, when a station wants to transmit, it listens to the transmission medium. When a node detects a carrier, its Carrier Sense is turned on and it will defer transmission until the medium is free: if two or more stations simultaneously begin to transmit, a collision occurs. In this case, the BEB algorithm for a random time interval is employed as below:

- When a collision occurs, each CSMA/CD unit chooses to back off for a period of time, determined by the maximal backoff value as an upper limit to the waiting period. Thus, this maximal backoff value is the number of time steps a node will wait at most before trying a retransmission. The first or initial backoff time value is termed ‘the minimal backoff time’. Thus, at the point of a first collision, the maximal back-off value is set to this minimal back-off time.

Each CSMA/CD unit will choose a random backoff time value which follows an equal distribution with an upper bound given by the maximal backoff value.

- If it should happen, that two nodes attempt to retransmit after backoff at the same time, the maximal back-off time value at each node involved in the collision is multiplied by 2 (maximum upper bound of 1024 for the factor). Thus, this causes an exponential growth of the maximal backoff time, i.e. binary exponential backoff.

- On a successful transmission, the transmitting unit sets its backoff value to zero.

- If a unit has attempted backoff 16 times due to collisions for transmitting the same packet, the BEB algorithm forces that unit to discard that packet. Furthermore, the backoff value of this unit is reset to zero, i.e any new backoff/retransmission attempt will be determined again by the minimal backoff time.

It is evident, that a Linear Backoff scheme is easily employed. In a Linear Backoff scheme, the increase of the maximum backoff window is linear, i.e. $n_f$, on each successive failure, with counter, $n_f$, in contrast to $2^{n_f}$ for the BEB scheme. In the next section, an overview of our novel Media Access Control (MAC) strategy is given.

B. The suggested MAC for half-duplex Ethernet

In our work [11], [12], we suggested three MAC principles for real-time control communication. The first principle is the synchronization signal. The second is the introduction of a time slot after the synchronization signal for transmission of each Ethernet packet. Such time slots might be the same for some transmission nodes, i.e. we are reducing with this measure collisions but may not avoid them. This is practical as we are not using high performance switches/router equipment, so that even specific unique time slots may not fully avoid collisions. However, we are not going to use this suggested MAC-principle of time-slots in our analysis to keep things simple. The final principle is the application of different minimal backoff times for each MAC-unit. All data packets have a length sufficiently small, so that they can be transmitted within a sufficiently small time interval. However, note that the analysis here is based on the basic idea of time steps (which have for instance a duration of 5.12 ms in a 100Mbit/s Ethernet network) and does not carry physical units. Thus, the two considered MAC-principles are:

1) Synchronization signal: Real-time applications require tight synchronization so that the delivery of control messages can be guaranteed within defined message cycle times. Practical implementation is possible by using the IEEE 1588 clock synchronization approach. Thus, a synchronization signal for data transmission is sent at the beginning of a transmission period, where all transmitting nodes are allowed to send a data packet.

2In our practical papers[11], [12], the length of one data packets was fixed to 48 bytes for a 100Mbit/s network. Fixed-length transmission has a predictable transmission time and reduces the probability of frame collision.
The time gap between each synchronization signal has to be sufficiently large (i.e. several time steps) to permit transmission of all data packets within the interval defined by the synchronization signal. The CSMA/CD methods have to guarantee that all packet transmissions following a single synchronization signals is carried out with minimal delay and jitter.

2) The backoff scheme (CSMA/CD) comes into play for the half duplex Ethernet network. Applying different, but fixed minimal backoff times of one or multiple time steps for each transmitting node will allow that some collisions are avoided early on and jitter and delay is kept small.

By assuming an equally distributed probability for collision, transmission and backoff (as explained later), we will now develop a rigorous Markov model for the suggested MAC scheme, which will allow us to show the advantage of the suggested scheme. For this, we provide a short introduction to Markov chains and our model principles first.

III. OVERVIEW OF MARKOV CHAIN MODELLING

Markov chains are an approach in mathematical modelling for random phenomena involving transitions between states with well defined probabilities: i.e. A.A.Markov, 1907 [16] explored chance processes. With this approach, the outcome of a test/experiment can affect the outcome of the next experiment, creating a Markovian chain.

In general, a Markov chain considers a process with state $X_n$ at time $n$, where the state $X_n$ is an integer between zero and $N$. The state constitutes of what is being modelled, and for the purpose of this overview is left abstract. The one-step transition probability is

$$ P(X_{n+1} = b | X_n = a) $$

This probability determines the chance that a process moves from state $X_n = a$, at time step $n$ to state $X_{n+1} = b$ at the later time step $n + 1$.

This allows to create large scale decision processes represented by graphs. Usually, the number of states is finite, i.e. creating a finite sized graph for a process of possibly infinite duration.

A state transition graph for $N = 3$ is shown in Figure 1. For any given state $a$, Equation (2) must hold;

$$ \sum_{k=0}^{N} P_{ak} = 1 $$

This states that the sum of probabilities $P_{ak} = P(b_k | a)$ of all possible $N$ successor states, $b_k$, $k = 1, ..., N$ from a single state, $a$ must be one.

Hence, a finite number of possible outcomes at each stage is required to make sure that we know the probabilities for any particular outcome at the $j$-th stage, given the knowledge of the outcome for the first $j - 1$ stages. For each $j$, a tree graph $U_j$ is obtained [17]. The set paths of this tree serve as a possibility space for any statements relating to the $j$ experiments. Figure 2 illustrates the general concept of our communication approach. Thus, the decision model of a Markov chain is a way of analyzing a set of probabilistic decisions that happen one after another.
is sufficient to describe the state of the transmission process. Here, \( n \) describes the transition step for the Markov chain, i.e. \( n = 0, 1, 2, 3, \ldots N \). Thus, the two node backoff process starts at step \( n = 0 \) and continues to step \( n = 1 \). \( BC_1 \) and \( BC_2 \) are the counters in the backoff scheme which count the number of failed attempts for transmission of data packets for each node. Thus, they determine the maximum waiting duration per node. For the Linear Backoff scheme, this is:

\[
(BC_1 \cdot BT_1 \text{ and } BC_2 \cdot BT_2)
\]

where \( BT_1 \) and \( BT_2 \) are each the minimal backoff times. For the Binary Exponential backoff scheme, the maximum waiting periods are given by:

\[
(BT_1 \cdot 2^{BC_1} \text{ and } BT_2 \cdot 2^{BC_2})
\]

These values will act as upper bounds for the parameters \( BD_1 \) and \( BD_2 \). Thus, the parameters \( BD_1 \) and \( BD_2 \) determine the current maximal backoff time values of the two nodes. They are chosen according to the maximum counter value, as outlined above. \( BD_1 \) and \( BD_2 \) are decreased once a node is waiting for the reattempt of transmission, i.e. a single decrement is made at each time step. Once \( BD_1 \) and \( BD_2 \) are zero the waiting/backoff of the relevant node is finished, i.e. this node will have to send data. Thus, \( BD_1 \) and \( BD_2 \) act as upper bound in which a node may wait or send data. Note that \( BC_1, BC_2, BD_1 \) and \( BD_2 \) are all free of units, i.e. integers.

This setup allows now to create a Markov chain tree for which the state parameter vector can be explained. As an example, Figure 3 shows the case where both nodes have identical minimal backoff times, \( BT_1 = 1 \) and \( BT_2 = 1 \) for the exponential backoff scheme. Note again that the model does not investigate the case of the sending in time slots, i.e. sending of a data packet occurs for all nodes at the occurrence of the synchronization signal followed by the backoff arbitration process if needed. Thus, as in Figure 3, the transmission process starts with:

\[
\{0, 0, 0, 1, 1\}
\]

This implies at this point \( BC_1 = BC_2 = 0 \). The backoff delay counters \( BD_1 = BD_2 = 1 \), will decrease by one which enforces the sending of a data packet. Thus, both nodes will make an attempt to send a data packet at the next step. This certainly will result in a collision, which forces both nodes to enter the backoff mechanism. This creates the state for the exponential backoff scheme:

\[
\{1, 1, 1, 2, 2\}
\]

At this point, \( BC_1 = BC_2 = 2 \) will result in the option that each node will either:

- Decrease the backoff counter, i.e. the next step will result in the state \( \{2,1,1,1,1\} \).
- Attempt successfully to send a data packet, i.e. \( BC_1, BD_1 \text{ or } BC_2, BD_2 \) are set to zero, while the other will
decrease its counter $BC_1$ or $BC_2$, i.e. $\{2.0,1.0,1\}$ and $\{2.1,0.1,0\}$.

- Cause a collision since both attempt packet transmission. This implies that both nodes will have to increase its attempt counter to $BC_1 = BC_2 = 2$. Thus, for the exponential backoff scheme the state will be $\{2,2,2,4,4\}$. Each of those options has the same probability of $\frac{1}{7}$.

The next steps follow a similar scheme. For instance, for $\{2.0,1.0,1\}$ and $\{2.1,0.1,0\}$: The transmission has been successful for one of the nodes. Thus, the backoff counter will decrease from 1 to zero for the other node in the third step and cause another successful transmission:

$\{3,0,0,0,0\}$

Here, the probability for each of the transitions is 1. Clearly, a complete graphical solution (see Figure 3 for slightly greater detail of the Markov tree) to this Binary Exponential Backoff model is not feasible, for which reason a Matlab-based computation can be conducted using the high performance computing facility of the University of Bristol. This will also enable the computation of jitter and delay as explained next.

A. Key characteristic values of the communication system: probabilities, jitter and delay

It is now possible to assess the probability of success and failure of transmission, which builds the basis for computation of jitter and delay. This is again done here for the case of the exponential backoff scheme with identical backoff time $BT_1 = BT_2 = 1$ (see Figure 3). Let us consider first, the probabilities of the initial value:

$$\begin{align*}
P_{x0} &= (x_0 = \{0,0,0,1,1\}) \\
&= 1
\end{align*}$$

This is followed by the transition probability:

$$\begin{align*}
P_{1,0,1} &= P(x_{1,1} = \{1,1,1,2,2\}|x_0 = \{0,0,0,1,1\}) \\
&= 1
\end{align*}$$

In step $n = 2$, the probability for observing a successful transmission, of either of the two nodes, a waiting cycle or a collision is equally distributed. Thus,

$$\begin{align*}
P_{2,i,4} &= P(x_{2,1,4}|x_{1,1} = \{1,1,1,2,2\}) \\
&= \frac{1}{4}
\end{align*}$$

where for our example:

$$\begin{align*}
x_{2,1,1} &= \{2,0,1,0,1\} \\
x_{2,1,2} &= \{2,1,0,1,0\} \\
x_{2,1,3} &= \{2,1,1,1,1\} \\
x_{2,1,4} &= \{2,2,2,4,4\}
\end{align*}$$

Thus, in general, for each transition probability of our chain we have:

$$\begin{align*}
P_{n,f,n} &= P(x_{n,a}|x_{n-1,f})
\end{align*}$$

where $n$ determines the step instance, $n = 0, 1, 2, 3, \ldots, N$, $f$ determines the state at step $(n-1)$ and $a$ determines the state, $a = 0, 1, 2, 3, \ldots, N$ at step $n$. It easily follows:

$$\begin{align*}
\sum_{a=0,1,2,\ldots,N} P_{n,f,a} &= 1
\end{align*}$$

for fixed value of $f$.

It is evident that the overall Markov chain is finite, i.e. each node will ultimately enter a state in which it was successful or unsuccessful in transmitting a data packet. Thus, those states which describe both nodes in the final state may be identified by the transition probabilities:

$$\begin{align*}
P_{u_1,f_1,a_1} 	ext{ and } P_{u_2,f_2,a_2}
\end{align*}$$

or

$$\begin{align*}
P_{u_1,f_1,a_1} 	ext{ and } P_{u_2,f_2,a_2}
\end{align*}$$

for Successful and Unsuccessful transmission of either node 1 or 2. There is only a finite number of triples $(n_{u_1}, f_{s_1}, a_{s_1}), (n_{u_2}, f_{s_2}, a_{s_2})$ and $(n_{u_2}, f_{s_2}, a_{s_2})$

This implies that the probabilities given by

$$\begin{align*}
P_{u_1} &= \sum_{(n_{u_1}, f_{u_1}, a_{u_1})} P_{n_{u_1}, f_{s_1}, a_{s_1}} \cdot P_{(n_{u_1}-1), f_{s_1}, \ldots \cdot P_{1,0,1}} \\
P_{u_2} &= \sum_{(n_{u_2}, f_{u_2}, a_{u_2})} P_{n_{u_2}, f_{s_2}, a_{s_2}} \cdot P_{(n_{u_2}-1), f_{s_2}, \ldots \cdot P_{1,0,1}}
\end{align*}$$

determines the probability that node 1 and node 2 send out each a packet with success. On the other hand, the probability

$$\begin{align*}
P_{u_1} &= \sum_{(n_{u_1}, f_{u_1}, a_{u_1})} P_{n_{u_1}, f_{u_1}, a_{u_1}} \cdot P_{(n_{u_1}-1), f_{s_1}, \ldots \cdot P_{1,0,1}}
\end{align*}$$

determines the probability of ultimately unsuccessful transmission. Moreover, $P_{s_1} = 1 - P_{u_1}, P_{s_2} = 1 - P_{u_2}$. This now also allows for the computation of the (average) delay of a successful transmission:

$$\begin{align*}
\tau_1 &= \sum_{(n_{u_1}, f_{u_1}, a_{u_1})} (n_{u_1} - 1) \cdot P_{n_{u_1}, f_{u_1}, a_{u_1}} \cdot P_{(n_{u_1}-1), f_{s_1}, \ldots \cdot P_{1,0,1}} \\
\tau_2 &= \sum_{(n_{u_2}, f_{u_2}, a_{u_2})} (n_{u_2} - 1) \cdot P_{n_{u_2}, f_{u_2}, a_{u_2}} \cdot P_{(n_{u_2}-1), f_{s_2}, \ldots \cdot P_{1,0,1}}
\end{align*}$$

for each node 1 and 2. The jitter is then defined based on the transmission delay $\tau_1$ and $\tau_2$ for each of the two nodes:

$$\begin{align*}
\sigma_j = \sqrt{\sum_{(n_{u_1}, f_{u_1}, a_{u_1})} (n_{u_1} - 1)^2} \cdot P_{n_{u_1}, f_{u_1}, a_{u_1}} \cdot P_{(n_{u_1}-1), f_{s_1}, \ldots \cdot P_{1,0,1}}
\end{align*}$$
\[ \sigma_{i2} = \sqrt{\sum_{n=1}^{N_{2}} (n_{s2} - \bar{f}_{2}) \cdot \sigma_{i2} \cdot P_{n_{s2}, f_{s2}, a_{s2}} \cdot P_{n_{s1}, f_{s1}, a_{s1}} \cdots P_{n_{s0}, f_{s0}}} \]

This allows now for assessment of delay, jitter and probability of transmission success for each of the nodes. The finite Markov state chain is for this reason computed to determine the three values \( \bar{\tau}_i, \sigma_i, P_{si} \) for \( i = 1, 2 \), i.e. for each node.

V. RESULTS

This section discusses performance analysis from the Markov model of the two-node transmitter system. In this case, a Matlab program is used to compute the network states as in Equation (3) of the Ethernet network and to evaluate the performance of the Ethernet communication approach.

A. Identical minimal backoff times

The scenario for Linear and Binary Exponential Backoff is computed, using in particular an overall retransmission attempt limit of 15 times and a minimal backoff time of 1 step for both transmission nodes. In this case, transmission jitter and delay are for both nodes identical. The result for the Linear Backoff scheme shows an advantage over the Binary Exponential Backoff scheme in terms of jitter and delay (Figure 4). This has been so far also confirmed in the simulation experiments of previous work [11], [12]. The performance improvement in terms of jitter and delay for the analytical result is about 25% for the Linear Backoff scheme in relation to the Binary Exponential approach.

B. Different minimal backoff times

In an attempt to enhance the existing standard MAC protocols, different backoff times have been applied for the nodes in the network system. Here we have chosen \( BT_1 = 1 \) and \( BT_2 = 2 \). Results in Figure 5 provide better performance in comparison to the case of identical backoff times.

Table I: Identical minimal backoff time for a two-node transmitter system.

<table>
<thead>
<tr>
<th>Scheme</th>
<th>Jitter Node 1</th>
<th>Jitter Node 2</th>
<th>Delay Node 1</th>
<th>Delay Node 2</th>
</tr>
</thead>
<tbody>
<tr>
<td>Binary Exponential Backoff</td>
<td>3.0158</td>
<td>3.0158</td>
<td>4.4854</td>
<td>4.4854</td>
</tr>
<tr>
<td>Linear Backoff</td>
<td>2.2365</td>
<td>2.2365</td>
<td>4.0277</td>
<td>4.0277</td>
</tr>
</tbody>
</table>
The node averaged delay for the Binary Exponential Backoff scheme has decreased by at least 15% in comparison to the identical backoff time scheme. In particular, jitter has decreased for both nodes in terms of the values for each node. The improvement of the averaged jitter for the Linear Backoff scheme is 26% while the improvement for the Binary Exponential Backoff scheme is 24%.

Obviously, the results demonstrate that the Ethernet MAC protocol performs reasonably better for the Linear Backoff scheme in comparison to the Binary Exponential Backoff scheme. Moreover, determinism increases by applying different minimal backoff times to each node.

VI. CONCLUSION

In this paper, the developed Markov chain model confirmed the observation from simulation and practical tests [11], [12] for a control communication strategy across half-duplex Ethernet. The two important principles for this communication scheme have been considered in this model: transmission at the synchronisation signal only and different minimal back-off time for each transmission node. A two node system has been considered in a setting of binary exponential back-off and linear back-off. These ideas compared to a standard minimal back-off time of 1 time step, have shown to provide a reduction of at least 24% in delay and jitter. It is evident that the introduction of time slots after the synchronisation event for transmission from each node (of which some time slots are assigned twice or more) will further reduce jitter and delay. Hence, this paper has theoretically confirmed recent practical and simulation results [11], [12] of these principal ideas of introducing a novel cheap time-triggered communication scheme in half-duplex Ethernet.

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